



Brief paper

Quantized identification of ARMA systems with colored measurement noise[☆]



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ABSTRACT

This paper studies the identification of ARMA systems with colored measurement noises using finite-level quantized observations. Compared with the case under colorless noises, this problem is more challenging. Our approach is to jointly design an adaptive quantizer and a recursive estimator to identify system parameters. Specifically, the quantizer uses the latest estimate to adjust its thresholds, and the estimator is updated by using quantized observations. To accommodate the temporal correlations of quantization errors and measurement noises, we construct a second-order statistics equivalent system, from which the original ARMA system is identified. The associated identifiability problem and convergence are analyzed as well. Finally, numerical simulations are performed to demonstrate the effectiveness of the proposed algorithm.

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1. Introduction

Quantized system identification is an important research topic, which aims to identify system parameters from quantized measurements rather than the exact measurements. Due to its practical applications, we have witnessed its tremendous development in the last decade. For instance, transmitting the coarsely quantized data in a networked system can improve the communication efficiency (Wang, Yin, Zhang, & Zhao, 2010) and storing quantized data reduces the memory size (Eldar & Kutyniok, 2012). While quantization is a severely nonlinear operator, it imposes great challenges in system identification.

To date, many quantized identification algorithms have been developed. They can be roughly categorized by the studied system models, e.g. gain system models (Li & Fang, 2007; Wang & Yin, 2007), FIR models (Godoy, Goodwin, Agueero, Marelli, & Wigren, 2011; Guo, Wang, Yin, Zhao, & Zhang, 2015; Guo & Zhao, 2013;

You, 2015; Yu, Zhang, & Xie, 2013), IIR models (Marelli, You, & Fu, 2013; Wang, Yin, & Zhang, 2006), time-varying systems (Bermudez & Bershady, 1996), and Hammerstein and Wiener models (Zhao, Wang, Yin, & Zhang, 2007; Zhao, Zhang, Wang, & Yin, 2010). On the other hand, they can also be classified based on the quantization setups, such as uniform or dithered quantizer (Geirhofer, Tong, & Sadler, 2006; Widrow & Kollar, 2008), fixed-level quantizer (Godoy et al., 2011; Marelli et al., 2013; Wimalajeewa & Varshney, 2012), binary quantizer (Guo & Zhao, 2013; Krishnamurthy & Poor, 1996; Vempaty, Ozdemir, Agrawal, Chen, & Varshney, 2013; Wang et al., 2006; Zhao et al., 2007), and adaptive quantizer (Bolcskei & Hlawatsch, 2001; Li & Fang, 2007; You, 2015). Compared with the static quantization, the adaptive version is more complicated but potentially more powerful, and may greatly reduce quantization effects on the identification accuracy. Hence, it has been intensively investigated in the literature.

In Fang and Li (2008), an adaptive quantized algorithm for distributed gain systems is proposed where the quantizer thresholds are dynamically adjusted from one sensor to another. This adjustment is conducted in the spatial domain and the estimation algorithm asymptotically approaches the Cramer–Rao lower bound (CRLB) as the number of sensors tends to infinity. Note that the measurement noises of each sensor is assumed to be spatially independent. In the time domain and under the maximum likelihood (ML) criterion, recursive quantized identification methods have been developed for FIR (Godoy et al., 2011)

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and ARMA systems (Marelli et al., 2013). Those recursive algorithms require to know the noise pdf in advance. To relax it, a recursive algorithm of the stochastic approximation type has been developed in You (2015) by jointly designing the quantizer and estimator. The adaptive quantizer uses the latest estimated parameters to tune its thresholds such that the quantizer operates like quantizing innovations.

However, all the aforementioned works deal with colorless noise models. There are only a few works to study the colored measurement noises, e.g., Mei, Wang, and Yin (2014), Wang and Yin (2010) where the noises are modeled as ϕ - and ρ -mixing processes, respectively. Since colored noises are common in practice, this is a meaningful problem. Along the same line, this paper focuses on developing a quantized recursive algorithm to identify the ARMA system and the AR/ARMA noise model, where the noise correlations will be exponentially decaying with respect to the time difference. While mixing types of correlated noises are broader than the colored case in this paper, the identification algorithms in Mei et al. (2014) and Wang and Yin (2010) are only applicable to *periodic* input signals. This is a fundamental assumption as they use an empirical-measure-based approach. Clearly, periodic input signal will limit the applicability of their quantized algorithms.

Inspired by You (2015), we jointly design the estimator and the quantizer in a unified framework. Particularly, the estimator provides the quantizer with the latest parameter estimate to adaptively adjust its thresholds. Such a strategy is motivated by the intuition that quantizing “innovations” is expected to be efficient. In this joint design scheme, the salient feature is that the estimator can recursively compute estimate of system parameters with the quantized observations and system inputs. Obviously, the system model on the estimator side has two correlated noise terms: one is the colored noises from the original system model and the other is the quantization errors, either of which makes it difficult to correctly identify the system parameters. To solve it, our idea is to construct an equivalent system with a hybrid noise term which has the same second-order statistics as the original system under quantized observations, and a recursive estimation algorithm is developed to identify the alternative system. It turns out that the alternative one is a standard Box–Jenkins model, whose parameters are estimated via the prediction-error method (Ljung, 1999). Based on this notion of equivalence, the unknown parameters of the original system can be estimated by using quantized observations. Moreover, this process can be implemented in a recursive way. Finally, the identifiability of the concerned problem is investigated and the convergence of the recursive algorithm is analyzed.

The rest of this paper is organized as follows. Section 2 formulates the quantized identification problem. Section 3 presents an identification method based on the joint design of the quantizer and estimator. Section 4 provides convergence analysis of the proposed identification algorithm. Section 5 extends the proposed method to the Box–Jenkins system model. In Section 6, simulation results are given to illustrate the performance of the developed identification method, followed by the conclusion in Section 7.

2. Problem formulation

We consider a networked ARMA system in Fig. 1 with measurement noises generated by an AR model:

$$\begin{aligned} y(t) &= \frac{B(q)}{A(q)}u(t) + \frac{1}{D(q)}e(t), \\ z(t) &= \mathcal{Q}_t[y(t)] \in \mathbb{R} \end{aligned} \quad (1)$$

where q denotes the forward shift operator. $u(t) \in \mathbb{R}$ and $y(t) \in \mathbb{R}$ are the system input and output, respectively. $e(t) \in \mathbb{R}$ is a white

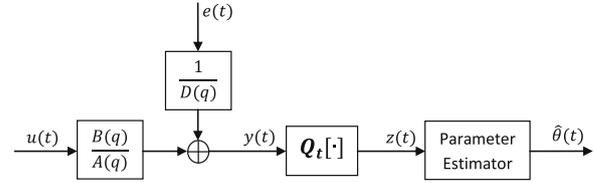


Fig. 1. System diagram.

Gaussian process, e.g., $e(t) \sim \mathcal{N}(0, \sigma_e^2)$, $z(t) \in \mathbb{R}$ is the quantized observation. Moreover, $A(q)$, $B(q)$ and $D(q)$ are defined by

$$\begin{aligned} A(q) &= 1 + a_1q^{-1} + \dots + a_{n_a}q^{-n_a}, \\ B(q) &= b_0 + b_1q^{-1} + \dots + b_{n_b}q^{-n_b}, \\ D(q) &= 1 + d_1q^{-1} + \dots + d_{n_d}q^{-n_d}. \end{aligned}$$

The time-varying K -level scalar quantizer $\mathcal{Q}_t[\cdot]$ is generically defined by

$$\mathcal{Q}_t[y(t)] = \begin{cases} v_{t,1} & b_{t,0} < y(t) \leq b_{t,1} \\ v_{t,2} & b_{t,1} < y(t) \leq b_{t,2} \\ \vdots & \\ v_{t,K} & b_{t,K-1} < y(t) \leq b_{t,K} \end{cases} \quad (2)$$

where $\{v_{t,k}\}_{k=1}^K$ are quantization levels, $\{b_{t,k}\}_{k=0}^K$ are quantization thresholds with $b_{t,0} = -\infty$ and $b_{t,K} = \infty$. The inverse of quantizer is defined by

$$\mathcal{Q}_t^{-1}[v_{t,i}] = (b_{t,i-1}, b_{t,i}], \quad i = 1, 2, \dots, K.$$

In the sequel, the system in (1) is abbreviated to ARARX model, and the following standard assumptions (Ljung, 1999) are made.

- A1: The input signal $u(t)$, which can be either deterministic or stochastic, is bounded and persistently exciting;
- A2: $B(q)$ and $A(q)$ are coprime, and the transfer function $B(q)/A(q)$ is stable;
- A3: The transfer function of the noise term $1/D(q)$ is stable.

In order to focus on the essence of quantizer design, the orders of system in (1) are assumed to be known. Otherwise, we can use a high-order ARX model for approximation, and subsequently reduce it to the structure of the system in (1) by using the model reduction techniques (Ljung, 1999, Chapter 10).

Let $Z_t = \{z(j) | j = 1, \dots, t\}$ be a set consisting of t quantized observations. U_t and Y_t are sets of t precise input and output samples, respectively. Collect the system parameter vector by

$$\theta = [a_1, \dots, a_{n_a}, b_0, \dots, b_{n_b}, d_1, \dots, d_{n_d}]^T$$

with the superscript T denoting the vector transpose. Let E be a mathematical expectation operator. Denote $\hat{y}(t|t-1, \theta) = E(y(t)|Z_{t-1}, U_{t-1}, \theta)$ the predictor (prediction model) for the system output at time t . Let $\hat{\theta}_t$ denote an estimate of θ based on t available samples and θ^* the true value of system parameters.

The problem of interest is to jointly design an adaptive quantizer and a recursive estimator for the parameter estimation of the ARARX model in (1).

3. Quantized identification of the ARARX model

In this section, the quantizer and estimator will be jointly designed for identification task.

3.1. Adaptive quantization scheme

The quantized output is a discrete function which has non-zero values only at finite points. Denote $\epsilon(t) = z(t) - y(t)$ the quantization error. Then, the system in (1) can be rewritten as

$$z(t) = y(t) + \epsilon(t) = \frac{B(q)}{A(q)}u(t) + \frac{1}{D(q)}e(t) + \epsilon(t). \quad (3)$$

Clearly, the main difficulties in identifying the above system are threefold: (a) the quantization noise $\epsilon(t)$ might be a colored noise with unknown statistical properties; (b) the noise $e(t)$ and $\epsilon(t)$ are correlated; (c) A nontrivial $D(q)$ renders the existing quantized algorithms inapplicable. If $D(q) = 1$, it reduces to the model in Marelli et al. (2013).

Obviously, the whiteness of $\epsilon(t)$ will substantially facilitate the design of the identification algorithm. To the best of our knowledge, two types of quantizers are workable: (a) Uniform quantizer with an appropriate dither (Widrow & Kollar, 2008); (b) Predictive quantizer (Gersho & Gray, 1991; You, 2015). The first approach is time-invariant and easy to implement but at the expense of infinite quantization levels. It does not make sense for the moderate rate (say one or two-bit). In the simulation, the identification performance of dither quantization will be illustrated.

The second approach is time-varying which shifts along with the prediction $\hat{y}(t|t-1, \hat{\theta}_{t-1})$ and yields the quantized output

$$z(t) = \mathcal{Q}_t[y(t)] \\ = \hat{y}(t|t-1, \hat{\theta}_{t-1}) + \mathcal{Q}[y(t) - \hat{y}(t|t-1, \hat{\theta}_{t-1})], \quad (4)$$

where $\mathcal{Q}[\cdot]$ is a finite-level Lloyd–Max quantizer (Max, 1960). It is noteworthy that from the system diagram in Fig. 1 the quantizer can directly access the exact system outputs while the estimator cannot.

For the ARARX model in (1), the output prediction can be explicitly written as:

$$\hat{y}(t|t-1, \hat{\theta}_{t-1}) = \frac{B(q, \hat{\theta}_{t-1})D(q, \hat{\theta}_{t-1})}{A(q, \hat{\theta}_{t-1})}u(t) \\ + (1 - D(q, \hat{\theta}_{t-1}))y(t). \quad (5)$$

Suppose that $\hat{\theta}_t \rightarrow \theta^*$ as $t \rightarrow \infty$. By (4), we can write the quantization error as follows:

$$\epsilon(t) = z(t) - y(t) \\ = \hat{y}(t|t-1, \hat{\theta}_{t-1}) - y(t) + \mathcal{Q}[y(t) - \hat{y}(t|t-1, \hat{\theta}_{t-1})] \\ \rightarrow \mathcal{Q}[e(t)] - e(t). \quad (6)$$

Under this case, the quantized error is indeed a white noise as long as the estimated system parameters are sufficiently close to their true values.

3.2. Recursive estimation method

In this subsection, we develop a quantized algorithm for the estimator in Fig. 1. To achieve this goal, an equivalent system having the same second-order statistics as (3) is provided, based on which a recursive estimation algorithm can be designed. To the best of our knowledge, this idea has never been exploited in the literature on quantized identification.

3.2.1. Second-order statistics equivalent model

The second term on the right-hand side of (3) is a colored noise, which is also correlated with the quantization error $\epsilon(t)$. Therefore,

using traditional methods by ignoring the quantization error may not be able to obtain unbiased estimates. To this end, we construct an alternative model with the same second-order statistics as that in (1):

$$z(t) = \frac{B(q)}{A(q)}u(t) + \frac{C(q)}{D(q)}\eta(t), \quad (7)$$

where $\eta(t)$ is a white noise with mean zero and variance σ_η^2 , and $C(q) = 1 + c_1q^{-1} + \dots + c_{n_c}q^{-n_c}$ satisfies the following equation:

$$\sigma_\eta^2 C(q)C(q^{-1}) = \sigma_e^2 + \rho\sigma_e\sigma_\epsilon D(q) + \rho\sigma_e\sigma_\epsilon D(q^{-1}) \\ + \sigma_\epsilon^2 D(q)D(q^{-1}), \quad (8)$$

where the unknown correlation coefficient is conceptually given by

$$\rho = \frac{\text{cov}(e(t), \epsilon(t))}{\sigma_e\sigma_\epsilon} \quad (9)$$

and σ_ϵ^2 is the variance of the quantization error. Note that the identification algorithm to be given later does not use the unknown coefficient ρ . That is, it does not cause any problem even we do not know ρ .

The above also implies that $C(q)\eta(t)$ has the same spectrum as that of $e(t) + D(q)\epsilon(t)$. As the alternative system in (7) is a standard Box–Jenkins model, its parameters can be estimated using the prediction-error method (Ljung, 1999). Moreover, both models share the same system parameters $\{A(q), B(q), D(q)\}$, and the second-order statistics. This motivates to use the estimated parameters $\{A(q), B(q), D(q)\}$ in (7) to the quantizer so that it adaptively adjusts its thresholds. Specifically, we use the quantized observation $z(t)$ from (1) to identify unknown parameters in (7) where we deliberately assume that $z(t)$ is generated from the model in (7). The estimated parameters of $\{A(q), B(q), D(q)\}$ are then used to construct a predictor $\hat{y}(t+1|t, \hat{\theta}_t)$, based on which a new quantized observation $z(t+1)$ is produced by using (4). Repeat the above process, the unknown parameters in (1) are identified. The remaining problem is how to identify (7) recursively by using $z(t)$.

3.2.2. Recursive estimation algorithm

As shown in Fig. 1, the estimator has to be updated once a new quantized sample is available. Hence, it is necessary to develop a recursive identification algorithm. We use ϑ to represent the parameter vector containing the coefficients of $A(q), B(q), C(q)$ and $D(q)$. Note that the parameter vector θ is contained in ϑ . The prediction-error criterion for estimation can be written as

$$\hat{\vartheta}_t = \arg \min_{\vartheta} V_t(Z_t, \vartheta),$$

$$V_t(Z_t, \vartheta) = \frac{1}{t} \sum_{j=1}^t \frac{1}{2} (z(j) - \hat{z}(j|j-1, \vartheta))^2, \quad (10)$$

where the predictor $\hat{z}(t|t-1, \vartheta)$ is defined by

$$\hat{z}(t|t-1, \vartheta) = \frac{D(q)B(q)}{C(q)A(q)}u(t) + \left(1 - \frac{D(q)}{C(q)}\right)z(t). \quad (11)$$

A recursive algorithm to resolve the above optimization problem is obtained as follows (Ljung, 1999):

$$\hat{\vartheta}_t = \hat{\vartheta}_{t-1} + \mu_t R^{-1}(t, \hat{\vartheta}_{t-1}) \psi(t, \hat{\vartheta}_{t-1}) \\ \times (z(t) - \hat{z}(t|t-1, \hat{\vartheta}_{t-1})) \\ := \hat{\vartheta}_{t-1} + \mu_t d_{t-1}, \quad (12)$$

$$R(t, \hat{\vartheta}_{t-1}) = \frac{1}{t} \sum_{j=1}^t \psi(j, \hat{\vartheta}_{j-1}) \psi^T(j, \hat{\vartheta}_{j-1})$$

Proof. It can be straightforwardly obtained by following Theorem 9.1 of Ljung (1999), and the details are omitted.

By Proposition 3, the CRLB for the estimation of ϑ is $\sigma_\eta^2 P^{-1}$. As $\psi(t, \vartheta^*)$ is expressed in terms of $z(t)$, P^{-1} is a matrix having complicated relations with σ_η^2 . Therefore, it is difficult to explicitly show the dependence of the CRLB on σ_η^2 .

Next, we show the quantization effects on the value of σ_η^2 . Suppose that $\mathcal{Q}[\cdot]$ is a fixed-level Lloyd–Max quantizer (Max, 1960). Let $\zeta(t) = \mathcal{Q}[e(t)]$ and $\epsilon(t) = \zeta(t) - e(t)$. Then, it has the following properties:

$$\begin{aligned} E(\zeta(t)\epsilon(t)) &= 0, \\ E(e^2(t)) &= E(\zeta^2(t)) + E(\epsilon^2(t)), \\ \text{cov}(e(t), \epsilon(t)) &= -E(\epsilon^2(t)). \end{aligned}$$

Inserting $\text{cov}(e(t), \epsilon(t))$ into (8) yields $\sigma_\eta^2 C(q)C(q^{-1}) = \sigma_e^2 - \sigma_e^2 D(q) - \sigma_e^2 D(q^{-1}) + \sigma_e^2 D(q)D(q^{-1})$. This implies that

$$\sigma_\eta^2 = \frac{\sigma_e^2 - (2 - \|\mathbf{d}\|^2)\sigma_e^2}{\|\mathbf{c}\|^2}, \quad (14)$$

where $\mathbf{c} = [1, c_1, \dots, c_{n_c}]'$ and $\mathbf{d} = [1, d_1, \dots, d_{n_d}]'$.

By (14), it is clear that when $\|\mathbf{d}\|^2 > 2$, the value of σ_η^2 increases along with σ_e^2 . It implies that σ_η^2 will be larger when the number of quantization levels becomes fewer. If $\|\mathbf{d}\|^2 < 2$, the value of σ_η^2 will be smaller as the number of quantization levels becomes fewer. This is an interesting phenomenon since it suggests that the measurement noise in the alternative model might not be proportional to the number of quantization levels.

As shown in (4), the quantizer plays two roles: one is to compute the predicted output $\hat{y}(t|t-1, \hat{\theta}_{t-1})$, and the other is to send the estimator the quantized observation. In practice, the quantizer can access the exact system output $y(t)$, it is reasonable to assume that quantizer has the knowledge of the exact predictor as shown in (5). As shown in Eqs. (4) and (6), when $\hat{\theta}_t \rightarrow \theta^*$, it has that

$$\begin{aligned} z(t) &= \hat{y}(t|t-1, \hat{\theta}_{t-1}) + \mathcal{Q}[y(t) - \hat{y}(t|t-1, \hat{\theta}_{t-1})] \\ &\rightarrow \hat{y}(t|t-1, \theta^*) - e(t) + \mathcal{Q}[e(t)]. \end{aligned} \quad (15)$$

In the above equation, since $e(t)$ is a white noise, the quantization error $\mathcal{Q}[e(t)] - e(t)$ is generically a white noise (Godoy et al., 2011). When $-e(t) + \mathcal{Q}[e(t)]$ is a white noise and under Assumptions A1–A3, it can be verified that the true parameter vector θ^* is indeed the unique solution for the identification of (15) or (1). The developed identification algorithm summarized in Section 3.3 provides a recursive estimation approach for the integrated model (15). At time step t , substituting the expressions of $z(t)$ in (4) and $\hat{z}(t|t-1, \vartheta)$ in (11), the parameter update in (12) can be rewritten as

$$\begin{aligned} \hat{\vartheta}_t &= \hat{\vartheta}_{t-1} + \mu_t \Psi(t, \hat{\vartheta}_{t-1}) \left(\frac{D(q, \hat{\vartheta}_{t-1})}{C(q, \hat{\vartheta}_{t-1})} \hat{y}(t|t-1, \hat{\vartheta}_{t-1}) \right. \\ &\quad + \frac{D(q, \hat{\vartheta}_{t-1})}{C(q, \hat{\vartheta}_{t-1})} \mathcal{Q}[y(t) - \hat{y}(t|t-1, \hat{\vartheta}_{t-1})] \\ &\quad \left. - \frac{D(q, \hat{\vartheta}_{t-1})B(q, \hat{\vartheta}_{t-1})}{C(q, \hat{\vartheta}_{t-1})A(q, \hat{\vartheta}_{t-1})} u(t) \right), \end{aligned} \quad (16)$$

where $\Psi(t, \hat{\vartheta}_{t-1}) = R^{-1}(t, \hat{\vartheta}_{t-1})\psi(t, \hat{\vartheta}_{t-1})$. Suppose that $D(q) = 1$. It follows from (8) that $C(q) = 1$. Then, (16) can be simplified as

$$\hat{\vartheta}_t = \hat{\vartheta}_{t-1} + \mu_t \Psi(t, \hat{\vartheta}_{t-1}) \mathcal{Q}[y(t) - \hat{y}(t|t-1, \hat{\vartheta}_{t-1})], \quad (17)$$

where $\Psi(t, \hat{\vartheta}_{t-1})$ depends on U_{t-1} rather than Z_{t-1} or Y_{t-1} . It is remarked that the above parameter update performs like a quantized LMS algorithm (Bermudez & Bershada, 1996) or a recursive estimator of stochastic approximation type (You, 2015). Thus, the recursive algorithm in this paper can be adapted for identifying the FIR model in You (2015) and the ARMA model in Marelli et al. (2013).

5. Identification of the Box–Jenkins model

The striking feature of the proposed algorithm is that we can easily generalize it to identify the Box–Jenkins model using quantized observations

$$\begin{aligned} y(t) &= \frac{B(q)}{A(q)}u(t) + \frac{F(q)}{D(q)}e(t) \\ z(t) &= \mathcal{Q}_t[y(t)]. \end{aligned} \quad (18)$$

Under the predictive quantization scheme and denoting the quantization error as $\epsilon(t) = z(t) - y(t)$, we obtain

$$z(t) = \frac{B(q)}{A(q)}u(t) + \frac{F(q)}{D(q)}e(t) + \epsilon(t). \quad (19)$$

The equivalent system model having the same second-order statistics is written as

$$z(t) = \frac{B(q)}{A(q)}u(t) + \frac{C(q)}{D(q)}\eta(t) \quad (20)$$

where $\eta(t)$ is a white noise and

$$C(q) = 1 + c_1 q^{-1} + \dots + c_{n_c} q^{-n_c}$$

satisfies the following equation

$$\begin{aligned} \sigma_\eta^2 C(q)C(q^{-1}) &= \sigma_e^2 F(q)F(q^{-1}) + \rho \sigma_e \sigma_e F(q)D(q^{-1}) \\ &\quad + \rho \sigma_e \sigma_e F(q^{-1})D(q) + \sigma_e^2 D(q)D(q^{-1}) \end{aligned} \quad (21)$$

with ρ being defined in (9).

Let ϕ_t be the parameter vector of the Box–Jenkins model at time t . The associated output prediction is

$$\begin{aligned} \hat{y}(t+1|t, \phi_t) &= \frac{D(q, \phi_t)B(q, \phi_t)}{F(q, \phi_t)A(q, \phi_t)}u(t+1) \\ &\quad + \frac{F(q, \phi_t) - D(q, \phi_t)}{F(q, \phi_t)}y(t+1), \end{aligned}$$

and the quantized observation is generated by

$$z(t+1) = \hat{y}(t+1|t, \phi_t) + \mathcal{Q}[y(t+1) - \hat{y}(t+1|t, \phi_t)].$$

For the ARARX system, only $A(q)$, $B(q)$ and $D(q)$ are to be estimated. However, for the Box–Jenkins model, we have to compute $F(q)$ using the spectrum equivalency equation in (21) and send to the quantizer. The following lemma gives a sufficient condition for the unique solution of $F(q)$.

Lemma 4. Suppose that $A(q)$, $B(q)$, $C(q)$ and $D(q)$ in (20) are available. Let σ_e^2 , σ_e^2 and $\text{cov}(e(t)\epsilon(t))$ be known as a priori knowledge. Then $F(q)$ can be uniquely determined if $\sigma_e F(q) + \rho \sigma_e D(q)$ is a minimum-phase function, i.e. the amplitudes of its roots are less than one.

Proof. Eq. (21) can be recast as

$$\begin{aligned} \sigma_\eta^2 C(q)C(q^{-1}) &= [\sigma_e F(q) + \rho \sigma_e D(q)][\sigma_e F(q^{-1}) \\ &\quad + \rho \sigma_e D(q^{-1})] + (1 - \rho^2)\sigma_e^2 D(q)D(q^{-1}). \end{aligned} \quad (22)$$

Table 1
Coefficients of the ARARX model and the Box–Jenkins model.

a_1	a_2	b_0	b_1	f_1	f_2	d_1	d_2
−0.2000	0.4421	0.7000	0.3000	0.2014	−0.2707	−0.4040	0.5649

In addition, the variance of $\eta(t)$ can be unbiasedly estimated by Ljung (1999, Lemma II.1):

$$\hat{\sigma}_\eta^2 = \frac{1}{t - n_\phi} \sum_{j=1}^t \left(z(j) - \hat{z}(j|j-1, \hat{\phi}_{j-1}) \right)^2 \quad (23)$$

where n_ϕ is the dimension of the parameter vector ϕ . After identifying the system model in (20), the value of

$$\sigma_\eta^2 C(q)C(q^{-1}) - (1 - \rho^2)\sigma_\epsilon^2 D(q)D(q^{-1})$$

in (22) can be computed off-line. Since $\sigma_\epsilon F(q) + \rho\sigma_\epsilon D(q)$ is minimum-phase, it can be uniquely obtained by minimum and maximum-phase factorization. As a result, $F(q)$ is determined.

Remark 2. For the Box–Jenkins model, the joint-design of the adaptive quantizer and recursive estimator can be obtained as in the previous section. Assume that $F(q)/D(q)$ is irreducible and minimum phase. By Lemma 4, the alternative system model (20) is identifiable. Different from the identification of the ARARX model, the knowledge of σ_ϵ^2 , σ_η^2 and $\text{cov}(e(t)\epsilon(t))$ here should be known in advance. For the system model in (18), if we set the system input to $u(t) = 0$ for all times, then it becomes a blind system identification problem. Thus, the proposed identification algorithm can solve the quantized blind identification problem with the input being a white noise.

6. Numerical simulation

In this section, simulation examples are provided to illustrate the effectiveness of the quantized identification algorithm. The identification performance under different quantization schemes will be illustrated.

The input signal $u(t)$ is generated by a truncated standard white Gaussian noise in the interval $[-3, 3]$. The noise $e(t)$ is generated as a standard white Gaussian noise, which is uncorrelated with the input signal $u(t)$. In this section, the recursive estimation starts from the 151st sample. To obtain its initial point, we collect the first 150 quantized samples by a zero-mean static Lloyd–Max quantizer, and calculate the minimizer of the prediction error criterion in (10).

The numerical simulations are based on the following ARARX model and the Box–Jenkins model

$$y(t) = \frac{b_0 + b_1 q^{-1}}{1 + a_1 q^{-1} + a_2 q^{-2}} u(t) + \frac{1}{1 + d_1 q^{-1} + d_2 q^{-2}} e(t) \quad (24)$$

$$y(t) = \frac{b_0 + b_1 q^{-1}}{1 + a_1 q^{-1} + a_2 q^{-2}} u(t) + \frac{1 + f_1 q^{-1} + f_2 q^{-2}}{1 + d_1 q^{-1} + d_2 q^{-2}} e(t).$$

The associated true system parameters are given in Table 1 which are identifiable by applying a two-bit Lloyd–Max quantizer. Implementing a two-bit Lloyd–Max quantizer, the values of σ_ϵ^2 and $\text{cov}(e(t)\epsilon(t))$ can be computed off-line.

The mean square error (MSE) criterion is adopted to evaluate the identification performance:

$$\text{MSE}_t = \frac{1}{T} \sum_{i=1}^T \|\hat{\theta}_t^{(i)} - \theta^*\|_2^2, \quad (25)$$

where t is the time step, T denotes the number of Monte–Carlo runs, and $\hat{\theta}_t^{(i)}$ is the i -th estimate of the system parameters at the time step t .

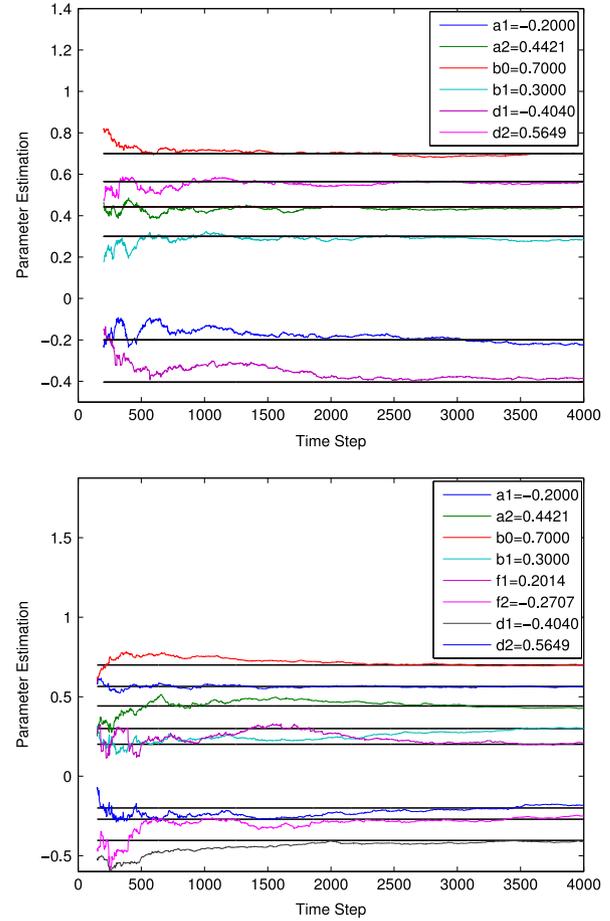


Fig. 2. Top: one sample trial of the ARARX model identification; bottom: one sample trial of the Box–Jenkins model identification.

In Fig. 2, the estimated parameters of the ARARX model and the Box–Jenkins model are plotted at different time steps by one Monte–Carlo trial. It can be observed that the estimated parameters fluctuate around their true values and the deviations become smaller as the number of quantized samples increases. Fig. 3 shows the estimate of the parameters by averaging 300 Monte–Carlo trials at each time step. The averaged estimates are close to their true values when the time index is large, which provides an experimental validation that the proposed identification works well under a two-bit Lloyd–Max quantizer. In addition, we observe that the MSE curve of the Box–Jenkins model decays slower than that of the ARARX model. This is mainly caused by the estimation of $F(q)$. Since the estimation of $F(q)$ is based on the estimated $C(q)$ and $D(q)$, the associated estimation error of $F(q)$ may be propagated and intensified from those of $C(q)$ and $D(q)$.

Next, we examine the identification performance of different quantization schemes based on the ARARX model in (24). Two alternative quantization schemes are adopted: a uniform quantizer with unit quantization interval and a two-bit static Lloyd–Max quantizer, and they all use the same identification algorithm which is developed in this paper. From Fig. 4, one can find that both the uniform quantizer and the adaptive quantizer can result in accurate estimates. For the uniform quantizer, it can be considered as a dithered quantizer since there already exists a white noise before quantization. However, the static Lloyd–Max quantizer leads to a biased estimation, which is caused by the fact that the quantization errors are temporally correlated. Moreover, Fig. 5 shows the identification performance of the adaptive Lloyd–Max quantizers with different numbers of quantization levels, where the convergence speed of the proposed identification algorithm

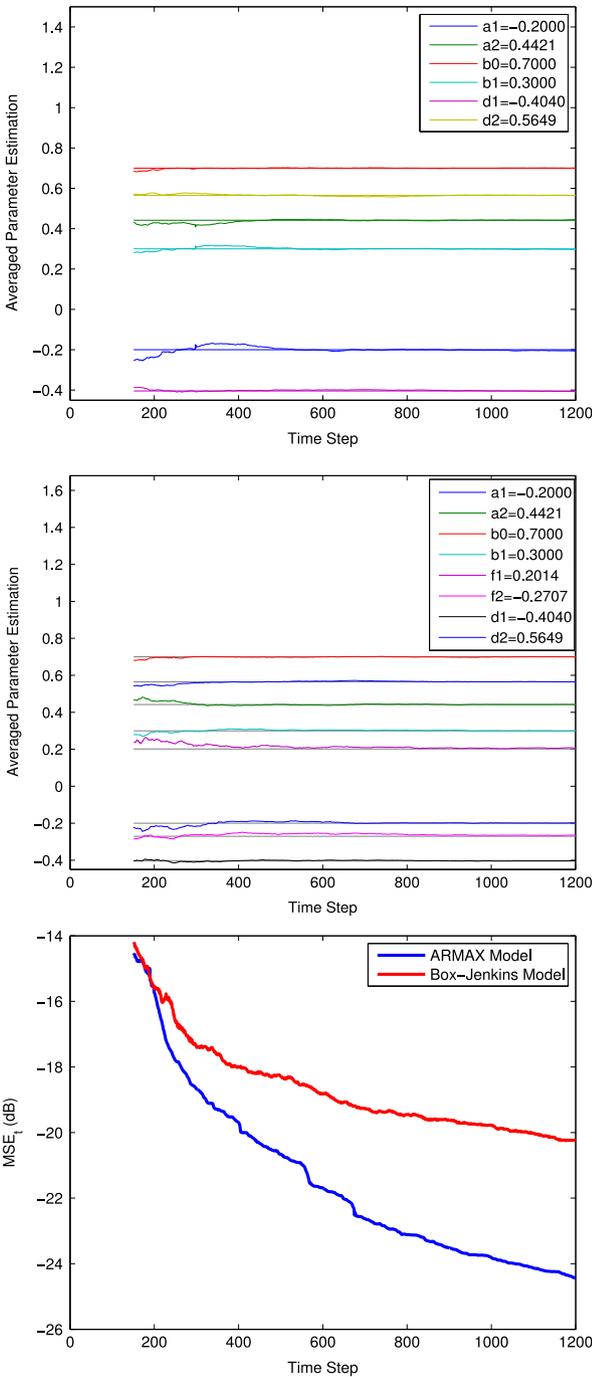


Fig. 3. Top: averaged identification result of the ARARX model; middle: averaged identification result of the Box-Jenkins model; bottom: MSE curves.

is much faster when more quantization levels are involved. It is noteworthy that the MSE can better reflect the performance of the proposed identification algorithm. The MSE values at the first few iterations may not be reliable, which are caused by following facts: (a) the initial conditions are randomly chosen; (b) the sequences of step sizes for different sample trials are distinct; (c) the associated recursive algorithm may not produce satisfactory results under a small number of observation samples.

7. Conclusion

In this paper, we have dealt with the quantized identification problems of the ARARX model and the Box-Jenkins model via

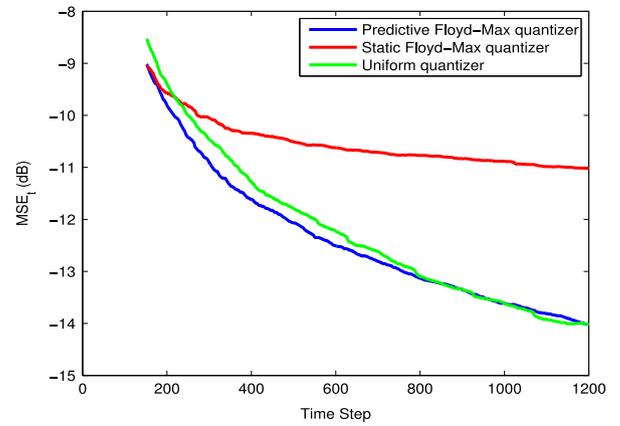


Fig. 4. MSE associated with the adaptive Lloyd-Max quantizer, static Lloyd-Max quantizer and uniform quantizer.

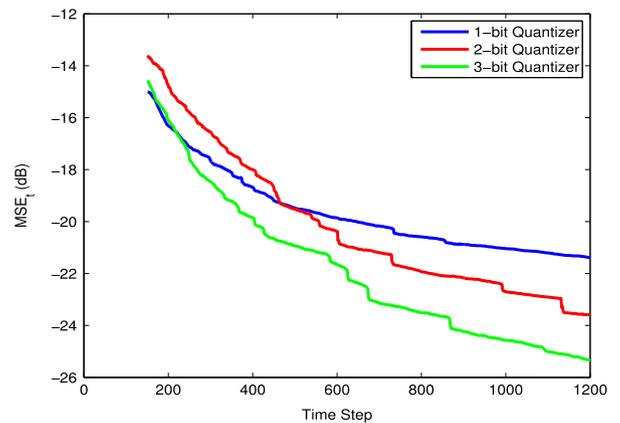


Fig. 5. MSE associated with the Lloyd-Max quantizers with different numbers of quantization levels.

jointly designing the quantizer and estimator. The designed quantizer adaptively adjusts its quantization thresholds according to the latest estimate of the system parameters, which aims to provide the estimator the “innovation” of outputs. For the estimator, it recursively estimates the system parameters based on the quantized observations. Since the received observations at the estimator are contaminated by the quantization error and colored measurement noise, a second-order statistically equivalent system model was constructed and identified. Simulation results show that the proposed method works well, even under one-bit quantized observations.

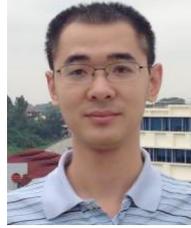
The connections between the existing adaptive quantized identification algorithms and the presented approach are discussed. It shows that the presented algorithm is a generalized version of quantized LMS algorithm or the recursive estimation of the stochastic approximation type. Thus, the presented algorithm can be applied for the identification of ARMA systems with white measurement noises. Moreover, when the concerned system model does not have the term involving the deterministic system input but the colored noise term, the associated quantized identification becomes a blind identification problem under quantized observations, which will be further investigated in our future work.

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